

## **REDUCED OUTPUT TOPOLOGY FOR MULTI-REFERENCE SWITCHING AMPLIFIERS**

### **REFERENCE TO RELATED APPLICATION**

This application claims priority to U.S. Provisional Patent Application Serial No. 60/406,207, filed August 27, 2002, the entire content of which is incorporated herein by reference.

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### **FIELD OF THE INVENTION**

This invention relates generally to multi-reference switching amplifiers and, in particular, to a simplified output topology associated with such amplifiers.

### **BACKGROUND OF THE INVENTION**

10 Multi-reference switching amplifiers of the type shown, for example, in PCT application PCT/US99/26691, entitled "Multi-Reference High Accuracy Switching Apparatus," yield significantly higher instantaneous resolution than standard switching amplifiers. The cost for this performance improvement, however, resides in an additional regulatory device and one or two switching devices (for non-bridged or bridged configurations, respectively) per reference added.

15 Particularly in cost-sensitive applications, there remains a need for a simplified output topology that retains the function and resolution inherent in multi-reference switching amplifiers.

### **SUMMARY OF THE INVENTION**

20 The present invention resides in a method and attendant circuitry for reducing the number of regulatory and switching devices in a multi-reference switching amplifier. In the preferred embodiment, multiple independently-modulated effective references are summed at a load through use of both linear and switched control of switching devices.

## BRIEF DESCRIPTION OF THE DRAWING

Figure 1 shows a preferred embodiment of the present invention.

## DETAILED DESCRIPTION OF THE INVENTION

Referring now to Figure 1, switching devices 125, 126, 127, and 128 form a  
5 bridged output known in the art as an "H" bridge. Inductors 129 and 130, in conjunction  
with capacitor 131, filter switching alias products from the load 132. Note that in this  
case only four output switching devices are used.

Data separator 101 isolates coarse data 102 and fine data 103 from incoming data  
stream 100. These data streams 102 and 103 are presented as inputs to pulswidth  
10 modulators 104 and 105, which proportionally convert said coarse data 102 and fine data  
103 into modulated coarse pulse stream 133 and fine pulse stream 134, respectively. If  
the sign 106 of the incoming data stream 100 is high, as indicated by data separator 101,  
switching device 125 is modulated by the coarse pulswidth stream 133, through AND  
gate 107. While the sign 106 is high, transmission gate 109 is activated, forcing the  
15 control input of switching device 126 to follow the complement of coarse pulse width  
stream 133, as inverted by inverter 121. Resistor 119 serves to limit output current of  
differential amplifier 111.

Conversely, if the indicated sign 106 of the incoming data stream 110 is low;  
switching devices 127 and 128 are modulated by the coarse pulswidth stream 133  
20 (through AND gate 108) and its complement (through transmission gate 110 and inverter  
121), respectively. Resistor 120 serves to limit output current of differential amplifier  
112. Coarse modulation in this fashion operates exactly as shown in the multi-reference  
application referenced above.

A second reference voltage, proportional to the power supply voltage  $V+$ , is  
25 formed by the resistor divider 123/124, and input to differential amplifiers 111 and 112.  
When not disturbed by transmission gate 109, switching device 125, or diode 115,  
differential amplifier 111 outputs a voltage to cause the output of switching device 126 to

equal the reference voltage formed by resistors 123 and 124. When the indicated sign 106 is low, NOR gate 113 turns on diode 115 with the inverse (from inverter 135) of the fine pulsewidth stream from pulsewidth modulator 105, forcing switching device 126 to turn on, through the resultant output increase of differential amplifier 111. This results in  
5 switching at the output of switching device 126 between ground and the reference voltage formed by resistors 123 and 124, inversely modulated by fine-resolution 103 provided to pulse width modulator 105.

When not disturbed by transmission gate 110, switching device 127, or diode 116, differential amplifier 112 outputs a voltage to cause the output of switching device 128 to  
10 equal the reference voltage formed by resistors 123 and 124. When the indicated sign 106 is high, NOR gate 114 turns on diode 116 with the inverse (from inverter 135) of the fine pulse width stream from pulse width modulator 105, forcing switching device 128 to turn on, through the resultant output increase of differential amplifier 112. This results in  
15 switching at the output of switching device 128 between ground and the reference voltage formed by resistors 123 and 124, inversely modulated by fine-resolution 103 provided to pulse width modulator 105.

In the discussion above, coarse-resolution data 102 is used to modulate V+ on one side of load 132, while fine-resolution data 103 is used to modulate the reference voltage formed by resistors 123 and 124 on the other side of load 132, under control of data sign  
20 106. Although summation at the load of multiple references, modulated by appropriate resolutions, directly follows the technique disclosed in the multi-reference application referenced above, note that this is accomplished by the present invention with significantly fewer output switching devices.

I claim: